

3.1 TWO TYPICAL INPUT SIGNALS USED IN ANALYSIS

LECTURE 6

UNIT IMPULSE FUNCTION

Consider the function $f_{\Delta}(x; a)$ shown in Fig. 1 and defined by Eqns. 1 and 2,

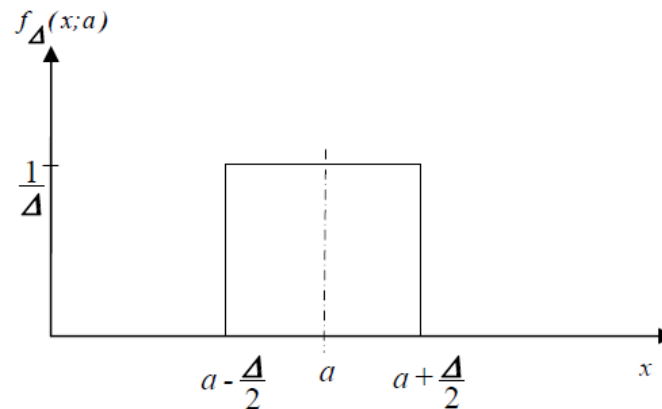
$$f_{\Delta}(x; a) = \begin{cases} 0 & -\infty < x < a - \frac{\Delta}{2} \\ \frac{1}{\Delta} & a - \frac{\Delta}{2} \leq x \leq a + \frac{\Delta}{2} \\ 0 & a + \frac{\Delta}{2} < x < \infty \end{cases} \quad (1)$$

The function has the unique property that

$$\int_{-\infty}^{\infty} f_{\Delta}(x; a) dx = 1 \quad (2)$$

Taking the limit of $f_{\Delta}(x; a)$, as $\Delta \rightarrow 0$ yields

$$\lim_{\Delta \rightarrow 0} f_{\Delta}(x; a) = \delta(x - a) = \begin{cases} 0 & x \neq a \\ \infty & x = a \end{cases} \quad (3)$$



From Eq. (2)

$$\int_{-\infty}^{\infty} \delta(x - a) dx = 1 \quad (4)$$

The function defined by Eq. (3) and (4) is called the unit impulse function (approximate forms are eqns. 1 and 2). It has many applications in physics and engineering. It is used to mathematically represent physical quantities that act for a short time, i.e., an impulsive force, and a concentrated load applied at a location $x = a$ on a structure.

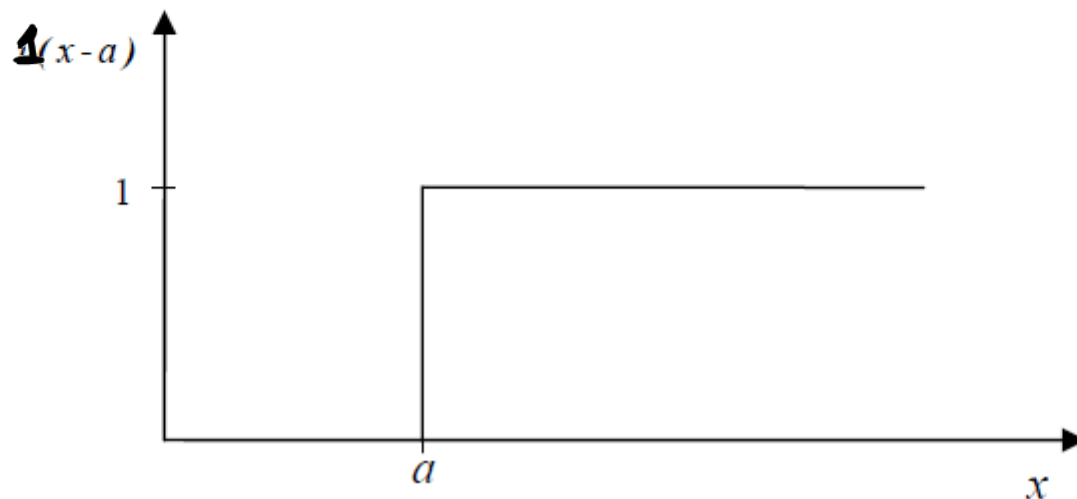
UNIT STEP FUNCTION : $\mathbf{1}(t) = \begin{cases} 0 & t < 0 \\ 1, & t \geq 0 \end{cases}$

Now define

$$\mathbf{1}(x-a) = \int_0^x \delta(x-a) dx = \int_0^x \lim_{\Delta \rightarrow 0} f_{\Delta}(x;a) dx = \begin{cases} 0 & x \leq a \\ 1 & x > a \end{cases} \quad (5)$$

The function defined in Eq. (5) is called the unit step function and is plotted in Fig. 2. Differentiating Eq. (5) gives

$$\frac{d\mathbf{1}(x-a)}{dx} = \delta(x-a) \quad (6)$$



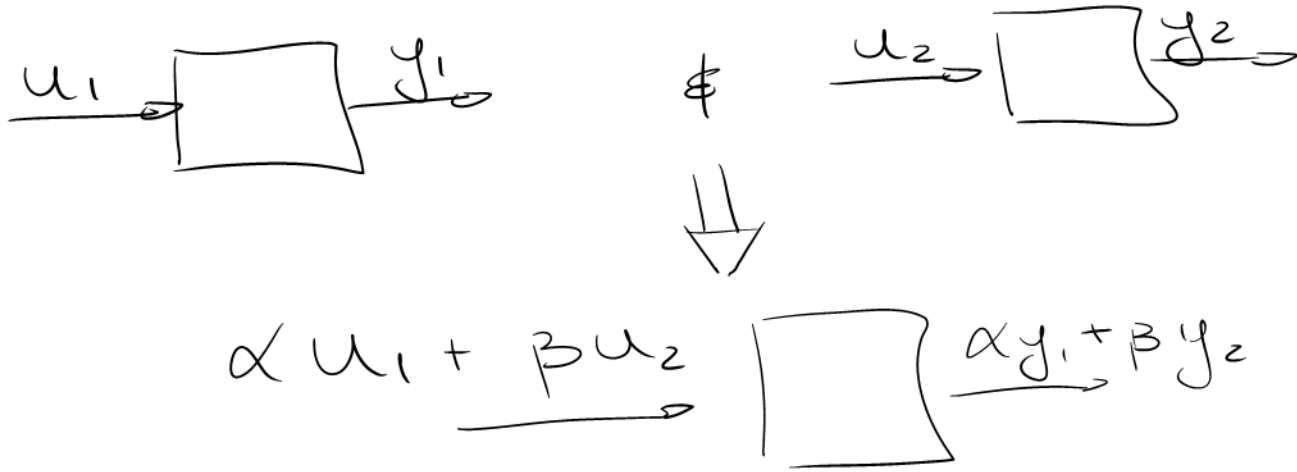
Note: You can integrate a step to get a ramp, and a ramp to get a parabola,and so on.

The definitions of the unit impulse function and unit step function can also be used to derive the following integral formulas. For any function ($g(t)$),

$$\int_0^t \delta(\tau - a)g(\tau)d\tau = \mathbf{1}(t - a)g(a) \quad (7)$$

$$\int_0^t \mathbf{1}(\tau - a)g(\tau)d\tau = \mathbf{1}(t - a)\int_a^t g(\tau)d\tau \quad (8)$$

SUPERPOSITION



Ex SHOW THAT SUPERPOSITION HOLDS FOR

$$\dot{y} + Ky = u$$

$$\dot{y}_1 + Ky_1 = u_1$$

$$\dot{y}_2 + Ky_2 = u_2$$

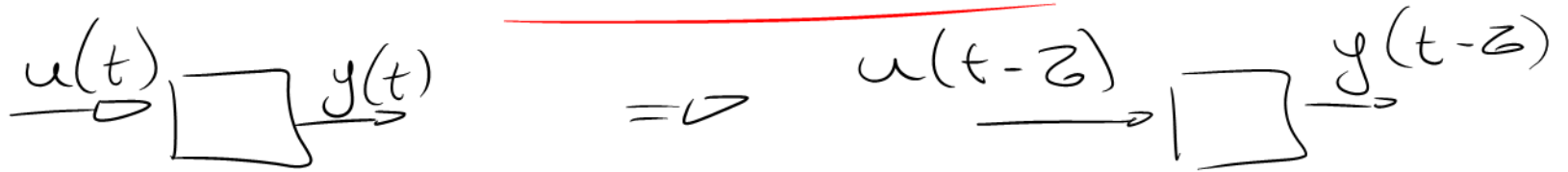
$$\Rightarrow \alpha (\dot{y}_1 + Ky_1) = \alpha u_1$$

$$\Rightarrow \beta (\dot{y}_2 + Ky_2) = \beta u_2$$

$$\alpha (\dot{y}_1 + Ky_1 - u_1) + \beta (\dot{y}_2 + Ky_2 - u_2) = 0$$

$$0 + 0 = 0 \quad \checkmark$$

TIME INVARIANCE



Ex SHOW THAT $\dot{y}(t) + \kappa(t)y(t) = u(t)$ is TI

$$\dot{y}_1(t) + \kappa(t)y_1(t) = u_1(t) \quad (1)$$

$$\dot{y}_2(t) + \kappa(t)y_2(t) = u_1(t-z)$$

$$\text{if } y_2(t) = y_1(t-z) \Rightarrow \dot{y}_1(t-z) + \kappa(t)y_1(t-z) = u_1(t-z)$$

$$\text{for } \eta = t-z \Rightarrow \dot{y}_1(\eta) + \kappa(\eta+z)y_1(\eta) = u_1(\eta)$$

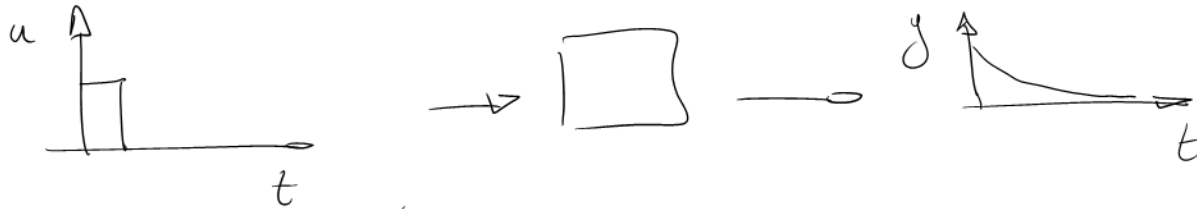
which is = (1) only if $z=0$ or

$$\kappa(\eta+z) = \kappa(\eta) \quad \forall z$$

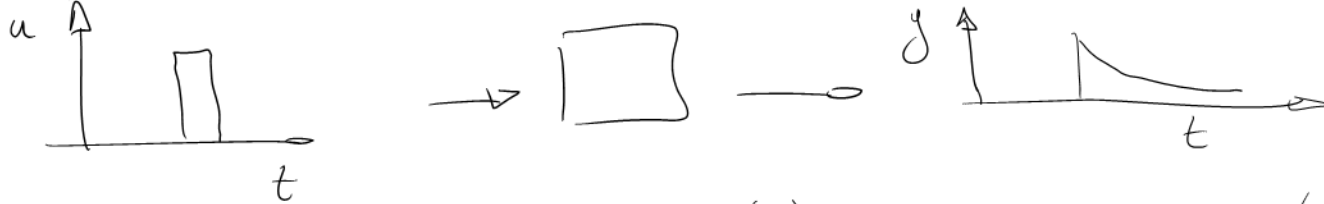
$$\Rightarrow \kappa = \text{cte}$$

IF WE PUT TOGETHER THESE FOUR CONCEPTS (IMPULSE, STEP, SUPERPOSITION & TIME INVARIANCE) ...

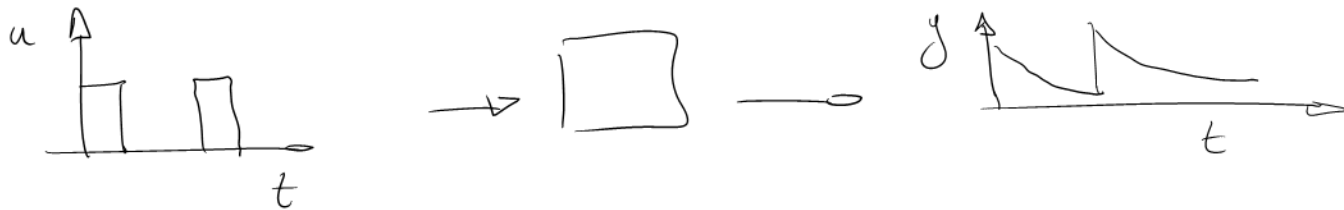
IMAGINE THAT FOR AN INPUT $u(t)$, OUR LTI SYSTEM RESPONDS w/ $y(t)$ AS SO:



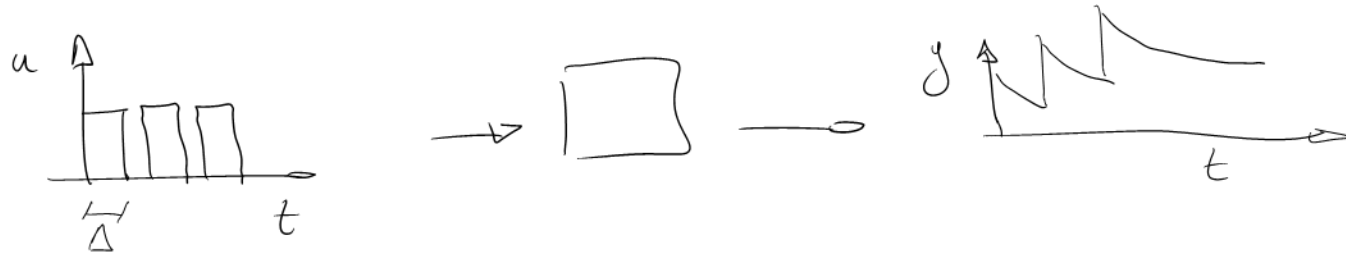
So, FOR $u(t-z)$, THE OUTPUT WOULD BE



By SUPERPOSITION, IF $u(t)$ IS THE TWO $u(t)$'s ABOVE:



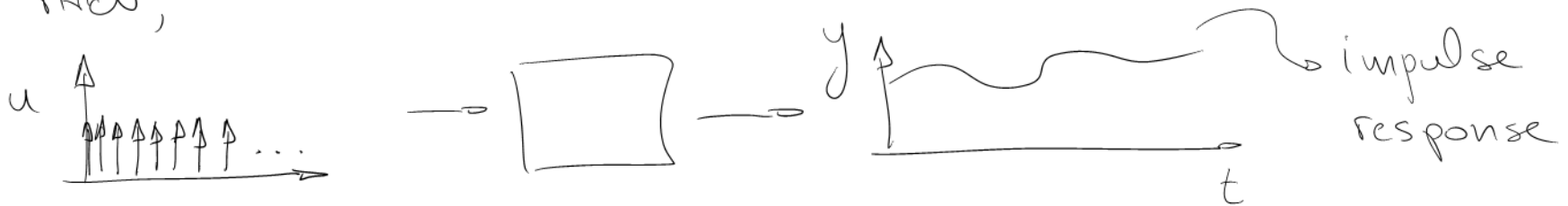
AND GENERALIZING THIS IDEA...



NOW, IF THE WIDTH OF THE PULSES $u(t)$ BECOME VERY NARROW ($\lim_{\Delta \rightarrow 0}$ THAT IS: $\delta(t)$ = UNIT IMPULSE FUNCTION)

AND THE DISTANCE BTW THEM ALSO VERY SMALL,

THEN,



FOR A CONTINUOUS FUNCTION $f(t)$:

$$\int_{-\infty}^{\infty} f(z) \delta(t-z) dz = f(t)$$

SO, IN THE CASE OF A LTI SYSTEM, WE

SAY THAT

$$\int_{-\infty}^{\infty} h(z) \delta(t-z) dz = y(t)$$

impulse response

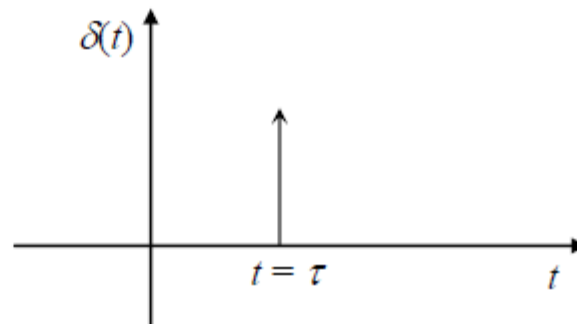
3.2 SUPERPOSITION AND CONVOLUTION INTERGRALS

- Dirac delta function or unit impulse function (mathematical fiction!)

$$\delta(t - \tau) = \begin{cases} \infty & t = \tau \\ 0 & t \neq \tau \end{cases}$$

and

$$\int_{-\infty}^{\infty} \delta(t - \tau) d\tau = 1$$



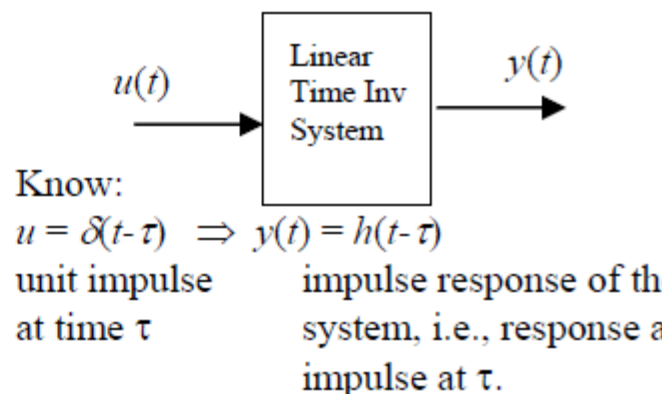
If $f(t)$ is continuous at $t = \tau$, this implies that

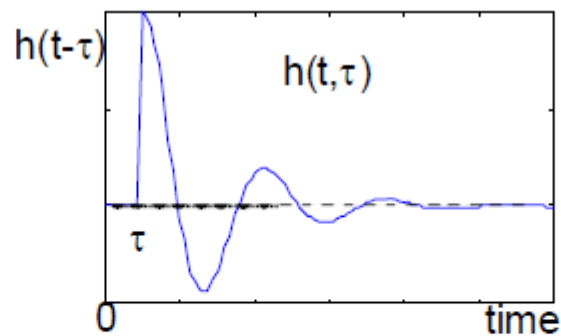
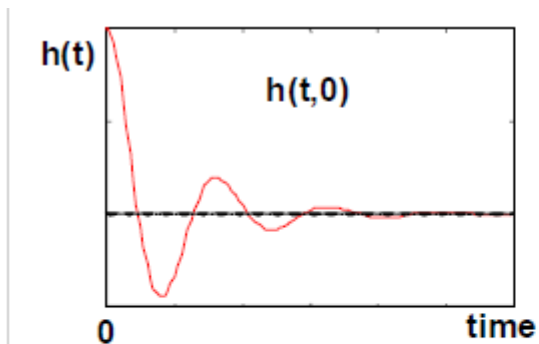
$$\int_{-\infty}^{\infty} f(\tau) \delta(t - \tau) d\tau = f(t)$$

- Impulse response: Response of a system to a unit impulse input. For a linear time invariant (constant coefficient) system, the impulse response is designated as $h(t)$ by convention. So, what is the response at t to an impulse applied at τ ? It is $h(t - \tau)$, i.e., it is just time shifted version of $h(t)$ (starts at τ instead of 0), as one might expect anyway! Now, the response to any input $u(t)$ can be obtained using superposition if $u(t)$ is broken up into lots of impulses! How do we do that? Can you imagine the time domain picture of $u(t)$ as a series of discrete impulses?

$$y(t) = \lim_{\Delta\tau \rightarrow 0} \sum_{\tau = -\infty}^{\infty} [u(\tau) d\tau] h(t - \tau) = \int_{-\infty}^{\infty} u(\tau) h(t - \tau) d\tau$$

This is the **superposition integral**.





The superposition integral provides the response for LTI systems, as

$$y(t) = \int_{-\infty}^{\infty} u(\tau)h(t-\tau)d\tau \quad \text{or} \quad y(t) = \int_{-\infty}^{\infty} u(t-\tau)h(\tau)d\tau$$

$\swarrow \eta = t - \tau$ \uparrow \eta = \tau

$$y(t) = \int_{-\infty}^{\infty} u(t-\eta)h(\eta)d\eta = \int_{-\infty}^{\infty} u(t-\eta)h(\eta)d\eta$$

This is called the **convolution integral**.

The convolution integral occurs so often that the notation $y = u * h$ is often used to represent the equation.

Utility: The response of the system to ANY time varying input can be determined using this convolution integral.

[For a general linear system, the impulse response can be represented as $h(t, \tau)$ and it is not equal to $h(t - \tau)$...with that change, all the development above holds.]

Ex

$$\ddot{y} + Ky = u = \delta(t)$$

$$y(0) = 0$$

$$\int_{0^-}^{0^+} (\ddot{y} + Ky) dt = \int_{0^-}^{0^+} \delta(t) dt$$

$$\int_{0^-}^{0^+} \ddot{y} dt + K \int_{0^-}^{0^+} y dt = 1$$

$$y|_{0^-}^{0^+} = 1 \Rightarrow y(0^+) - y(0^-) = 1$$

Assume $y = Ae^{st} \Rightarrow \dot{y} = sAe^{st}$

$$Ase^{st} + KAe^{st} = 0 \Rightarrow s + K = 0$$

$$s = -K$$

$$y(0^+) = 1 \quad \& \quad y(t) = A e^{st} \quad \Rightarrow \quad A = 1$$

$$y(t) \triangleq h(t) = e^{-\kappa t} \quad \text{since } h(t) \not\equiv \text{for } t < 0$$

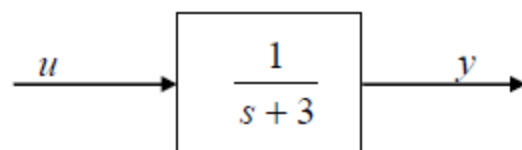
$$h(t) = e^{-\kappa t} * 1(t)$$

$$1(t) = \begin{cases} 0 & t < 0 \\ 1 & t \geq 0 \end{cases}$$

$$y(t) = \int_{-\infty}^{\infty} u(\tau) h(t-\tau) d\tau \quad \text{or} \quad y(t) = \int_{-\infty}^{\infty} u(t-\tau) h(\tau) d\tau$$

$$y(t) = \int_{-\infty}^{\infty} \delta(t-\tau) e^{-\kappa \tau} 1(\tau) d\tau =$$
$$= \int_0^{\infty} \delta(t-\tau) e^{-\kappa \tau} d\tau$$

Example: Given the following system:



Find the response y if u is a unit step?

Solution: The unit step input u can be expressed as

$$u(\tau) = \begin{cases} 1 & \tau \geq 0 \\ 0 & \tau < 0 \end{cases}$$

The impulse response of this system is given by

$$h = e^{-3t} \text{ (simply the inverse Laplace transform of the transfer function, since } L(\text{impulse})=1)$$

Since the system is linear and time invariant, $h(t, \tau) = h(t - \tau)$.

For this problem, $h(t - \tau) = e^{-3(t - \tau)}$.

Now perform the convolution operation to get the output y

$$\begin{aligned}y(t) &= \int_{-\infty}^{\infty} u(\tau)h(t-\tau)d\tau \\ &= \int_0^t 1 \cdot e^{-3(t-\tau)}d\tau \\ &= \frac{1}{3}(1-e^{-3t})\end{aligned}$$

Note that when $t < \tau$, $h(t-\tau) = 0$.

MATLAB Commands:

```
syms s t tau
G=1/(s+3);
u=1;
h=ilaplace(G);
h_tau=subs(h,t,t-tau);
y=int(h_tau*u,tau,0,t)
```

LAPLACE TRANSFORM

$$Y(s) = \int_{-\infty}^{\infty} y(t) e^{-st} dt$$

but if $y(t) = \int_{-\infty}^{\infty} h(t) u(t-z) dz$

$$Y(s) = \int_{-\infty}^{\infty} \left[\int_{-\infty}^{\infty} h(t) u(t-z) dz \right] e^{-st} dt$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \left[u(t-z) e^{-st} dt \right] h(z) dz$$

$$Y(s) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \left[u(\eta) e^{-s(\eta+z)} d\eta \right] h(z) dz$$

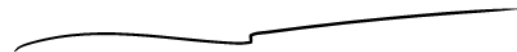
where $\eta = t - z$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} u(\eta) e^{-s\eta} d\eta e^{-sz} h(z) dz$$

$$= \int_{-\infty}^{\infty} u(\eta) e^{-s\eta} d\eta \int_{-\infty}^{\infty} e^{-sz} h(z) dz$$



$U(s)$



$H(s)$

$$Y(s) = U(s) * H(s)$$